

Plane waves in lossy media with $\sigma_2, \epsilon_2, \mu_2$.

$$\begin{aligned}\nabla \times \hat{\underline{H}}_2 &= \hat{\underline{J}}_2 + j\omega\epsilon_2\hat{\underline{E}}_2 = \sigma_2\hat{\underline{E}}_2 + j\omega\epsilon_2\hat{\underline{E}}_2 = (\sigma_2 + j\omega\epsilon_2)\hat{\underline{E}}_2 = j\omega\epsilon_2\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)\hat{\underline{E}}_2 \\ &= j\omega\hat{\epsilon}_{\text{eff}}\hat{\underline{E}}_2 \quad \hat{\epsilon}_{\text{eff}} = \epsilon_2\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)\end{aligned}$$

$$\hat{\epsilon}_{\text{eff}} = \epsilon_2\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)$$

$$\begin{aligned}\nabla \times \nabla \times \hat{\underline{E}}_2 &= -j\omega\mu_2\nabla \times \hat{\underline{H}}_2 \\ \nabla(\nabla \cdot \hat{\underline{E}}_2) - \nabla^2\hat{\underline{E}}_2 &= -j\omega\mu_2(j\omega\hat{\epsilon}_{\text{eff}}\hat{\underline{E}}_2), \quad \nabla \cdot \hat{\underline{E}}_2 = 0 \text{ for source-free medium} \\ \nabla^2\hat{\underline{E}}_2 + \omega^2\mu_2\hat{\epsilon}_{\text{eff}}\hat{\underline{E}}_2 &= 0 \\ \nabla^2\hat{\underline{E}}_2 + \hat{k}^2\hat{\underline{E}}_2 &= 0\end{aligned}$$

$$\hat{k} = \omega\sqrt{\mu_2\hat{\epsilon}_{\text{eff}}} = \omega\sqrt{\mu_2\epsilon_2\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)} = \omega\sqrt{\mu_2\epsilon_2}\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)^{1/2} \quad [m^{-1}]$$

$$\begin{aligned}\gamma &= j\hat{k} = j\omega\sqrt{\mu_2\hat{\epsilon}_{\text{eff}}} = j\omega\sqrt{\mu_2\epsilon_2\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)} \\ &= j\omega\sqrt{\mu_2\epsilon_2}\left(1 - j\frac{\sigma_2}{\omega\epsilon_2}\right)^{1/2} = \alpha + j\beta\end{aligned}$$

$$\text{loss} \Rightarrow \gamma \text{ is complex} \Rightarrow \gamma = \alpha + j\beta \Rightarrow k = \beta - j\alpha$$

if uniform plane wave is propagating in the \underline{k} direction

$$\begin{aligned}\hat{\underline{E}}(\underline{r}) &= \hat{\underline{E}}(x, y, z) = \hat{E}_x\underline{a}_x + \hat{E}_y\underline{a}_y + \hat{E}_z\underline{a}_z \\ &= \left[\hat{E}_{0x}\underline{a}_x + \hat{E}_{0y}\underline{a}_y + \hat{E}_{0z}\underline{a}_z\right]e^{-j(k_x x + k_y y + k_z z)} \quad \hat{\underline{E}}_0 \text{ is a constant vector} \\ &= \hat{\underline{E}}_0 e^{-j\hat{k}\cdot\underline{r}} = \hat{\underline{E}}_0 e^{-\gamma\underline{r}}\end{aligned}$$

$$\hat{\underline{E}}(\underline{r}, t) = \hat{\underline{E}}(x, y, z)e^{j\omega t} = \hat{\underline{E}}_0 e^{j\omega t} e^{-\gamma\underline{r}}$$

Perpendicular polarization

Oblique incidence at a dielectric boundary

Incident wave	Reflected wave	Refracted wave
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$$\begin{aligned}\hat{\underline{E}}_i &= \hat{E}_{i0}e^{-jk_i z}\underline{a}_y & \hat{\underline{E}}_r &= \hat{E}_{r0}e^{-jk_r z}\underline{a}_y & \hat{\underline{E}}_t &= \hat{E}_{t0}e^{-jk_t z}\underline{a}_y = \hat{E}_{t0}e^{-\gamma_t z}\underline{a}_y \\ \hat{\underline{H}}_i &= \frac{\hat{E}_{i0}e^{-jk_i z}}{\eta_1}(\underline{a}_{ni} \times \underline{a}_y) & \hat{\underline{H}}_r &= \frac{\hat{E}_{r0}e^{-jk_r z}}{\eta_1}(\underline{a}_{nr} \times \underline{a}_y) & \hat{\underline{H}}_t &= \frac{\hat{E}_{t0}e^{-jk_t z}}{\eta_2}(\underline{a}_{nt} \times \underline{a}_y)\end{aligned}$$

$$\begin{aligned} \underline{k}_i &= k_i \sin \theta_i \underline{a}_x + k_i \cos \theta_i \underline{a}_z & \Rightarrow & & \underline{k}_i &= \beta_1 \sin \theta_i \underline{a}_x + \beta_i \cos \theta_i \underline{a}_z, & \beta_1 &= k_i = k_r \\ \underline{k}_r &= k_r \sin \theta_r \underline{a}_x - k_r \cos \theta_r \underline{a}_z & & & \underline{k}_r &= \beta_1 \sin \theta_r \underline{a}_x - \beta_1 \cos \theta_r \underline{a}_z, & & \\ \underline{k}_t &= k_t \sin \theta_t \underline{a}_x + k_t \cos \theta_t \underline{a}_z = \gamma \sin \psi \underline{a}_x + \gamma \cos \psi \underline{a}_z, & & & \gamma_2 &= jk_t = \alpha_2 + j\beta_2, \psi = \theta_t \end{aligned}$$

$$\begin{cases} \hat{\underline{E}}_i = \hat{\underline{E}}_{i0} e^{-j(\beta_1 \sin \theta_i x + \beta_i \cos \theta_i z)} \underline{a}_y \\ \hat{\underline{E}}_r = \hat{\underline{E}}_{r0} e^{-j(\beta_1 \sin \theta_r x - \beta_1 \cos \theta_r z)} \underline{a}_y \\ \hat{\underline{E}}_t = \hat{\underline{E}}_{t0} e^{-(\gamma_2 \sin \psi x + \gamma_2 \cos \psi z)} \underline{a}_y \end{cases}$$

at the boundary $z = 0$,

$\hat{\underline{E}}$ tangential component continues

\Rightarrow

$$\underline{a}_y \cdot \hat{\underline{E}}_i|_{z=0} + \underline{a}_y \cdot \hat{\underline{E}}_r|_{z=0} = \underline{a}_y \cdot \hat{\underline{E}}_t|_{z=0}$$

$$\begin{aligned} & \hat{\underline{E}}_{i0} e^{-j\beta_1 \sin \theta_i x} \underline{a}_y + \hat{\underline{E}}_{r0} e^{-j\beta_1 \sin \theta_r x} \underline{a}_y \\ &= \hat{\underline{E}}_{t0} e^{-\gamma_2 \sin \psi x} \underline{a}_y \end{aligned}$$

$$\hat{\underline{E}}_{i0} + \hat{\underline{E}}_{r0} = \hat{\underline{E}}_{t0}$$

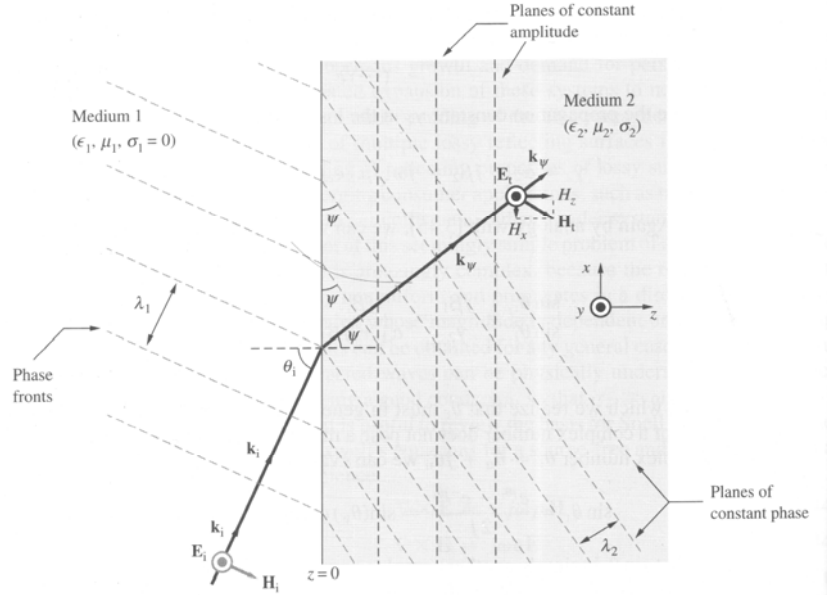


FIGURE 3.46. Oblique incidence on a lossy medium: Constant amplitude and phase fronts. The planes of constant amplitude are parallel to the interface, while the planes of constant phase are perpendicular to \underline{k}_ψ , which is at an angle ψ from the vertical. This "true" angle ψ is given by [3.59] and is also equal to the real part of the complex angle θ_t defined by [3.55].

$$\begin{aligned} \beta_1 \sin \theta_i x &= \beta_1 \sin \theta_r x & \Rightarrow & & \theta_i &= \theta_r \\ j\beta_1 \sin \theta_i x &= \gamma_2 \sin \psi x & & & j\beta_1 \sin \theta_i &= \gamma_2 \sin \psi \end{aligned}$$

$$\frac{\sin \psi}{\sin \theta_i} = \frac{j\beta_1}{\gamma_2} = \frac{j\beta_1}{\alpha_2 + j\beta_2}$$

$$\gamma_2 \cos \psi = \gamma_2 \sqrt{1 - \sin^2 \psi} = \sqrt{\gamma_2^2 - \gamma_2^2 \sin^2 \psi} = \sqrt{\gamma_2^2 + \beta_1^2 \sin^2 \theta_i} \equiv p + iq$$

$$p \equiv \text{Re} \left\{ \sqrt{\gamma_2^2 + \beta_1^2 \sin^2 \theta_i} \right\} \quad q \equiv \text{Im} \left\{ \sqrt{\gamma_2^2 + \beta_1^2 \sin^2 \theta_i} \right\}$$

$$\hat{\underline{E}}_t = \hat{\underline{E}}_{t0} e^{-(\gamma_2 \sin \psi x + \gamma_2 \cos \psi z)} \underline{a}_y = \hat{\underline{E}}_{t0} e^{-p} e^{-j(\beta_1 \sin \theta_i x + qz)} \underline{a}_y$$

$$\beta_1 \sin \theta_i x + qz = (\beta_1 \sin \theta_i \underline{a}_x + q \underline{a}_z) \cdot (x \underline{a}_x + z \underline{a}_z) = \underline{k}_\psi \cdot \underline{r}$$

$$\tan \psi = \frac{\beta_1 \sin \theta_i}{q}$$

$$\begin{cases} \text{attenuation constant } \alpha = \omega \sqrt{\frac{\mu \epsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \epsilon} \right)^2} - 1 \right]^{1/2} & [np/m] \\ \text{phase constant } \beta = \omega \sqrt{\frac{\mu \epsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \epsilon} \right)^2} + 1 \right]^{1/2} & [rad/m] \end{cases}$$

prove:

$$\hat{k} = \sqrt{-j\omega\mu(\sigma + j\omega\varepsilon)} = \omega\sqrt{\mu\varepsilon\left(1 - j\frac{\sigma}{\omega\varepsilon}\right)}$$

$$e^{-j\theta} = \cos\theta - j\sin\theta = \cos\theta(1 - j\tan\theta) = \cos\theta(1 - jx)$$

$$x = \tan\theta \Rightarrow \begin{cases} \sin\theta = \frac{x}{\sqrt{x^2+1}} \\ \cos\theta = \frac{1}{\sqrt{x^2+1}} \end{cases} \quad \begin{cases} \cos(\theta/2) = \sqrt{\frac{1+\cos\theta}{2}} \\ \sin(\theta/2) = \sqrt{\frac{1-\cos\theta}{2}} \end{cases}$$

$$1 - jx = \frac{e^{-j\theta}}{\cos\theta} = 1 - j\frac{\sigma}{\omega\varepsilon}$$

$$\begin{aligned} (1 - jx)^{1/2} &= \frac{e^{-j\theta/2}}{(\cos\theta)^{1/2}} = \frac{\cos(\theta/2) - j\sin(\theta/2)}{(\cos\theta)^{1/2}} = \frac{\sqrt{\frac{1+\cos\theta}{2}} - j\sqrt{\frac{1-\cos\theta}{2}}}{(\cos\theta)^{1/2}} \\ &= \frac{\left(\frac{1 + \frac{1}{\sqrt{x^2+1}}}{2}\right)^{1/2} - j\left(\frac{1 - \frac{1}{\sqrt{x^2+1}}}{2}\right)^{1/2}}{\left(\frac{1}{\sqrt{x^2+1}}\right)^{1/2}} = \frac{1}{\sqrt{2}} \left[\left(\sqrt{x^2+1} + 1\right)^{1/2} - j\left(\sqrt{x^2+1} - 1\right)^{1/2} \right] \\ &= \frac{1}{\sqrt{2}} \left\{ \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1 \right]^{1/2} - j \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1 \right]^{1/2} \right\} \end{aligned}$$

$$\begin{aligned} \hat{k} &= \omega\sqrt{\mu\varepsilon\left(1 - j\frac{\sigma}{\omega\varepsilon}\right)} = \omega\sqrt{\mu\varepsilon} \cdot \frac{1}{\sqrt{2}} \left\{ \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1 \right]^{1/2} - j \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1 \right]^{1/2} \right\} \\ &= \omega\sqrt{\frac{\mu\varepsilon}{2}} \left\{ \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1 \right]^{1/2} - j \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1 \right]^{1/2} \right\} \end{aligned}$$

$$\gamma = j\hat{k} = j\omega\sqrt{\mu\varepsilon} = j\omega\sqrt{\mu\varepsilon\left(1 - j\frac{\sigma}{\omega\varepsilon}\right)}$$

$$= j\omega\sqrt{\frac{\mu\varepsilon}{2}} \left\{ \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1 \right]^{1/2} - j \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1 \right]^{1/2} \right\}$$

$$= \omega\sqrt{\frac{\mu\varepsilon}{2}} \left\{ \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1 \right]^{1/2} + j \left[\sqrt{1 + \left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1 \right]^{1/2} \right\} = \alpha + j\beta$$

$$\begin{cases} \alpha = \omega \sqrt{\frac{\mu \varepsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^2} - 1 \right]^{1/2} \\ \beta = \omega \sqrt{\frac{\mu \varepsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^2} + 1 \right]^{1/2} \end{cases}$$

complex intrinsic impedance of a conducting medium

$$\hat{\eta}_c = |\hat{\eta}_c| e^{j\phi_\eta} = \sqrt{\frac{\mu}{\hat{\varepsilon}}} = \sqrt{\frac{\mu}{\varepsilon \left(1 - j \frac{\sigma}{\omega \varepsilon}\right)}} = \frac{\sqrt{\frac{\mu}{\varepsilon}}}{\left[1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^2\right]^{1/4}} e^{j(1/2)\tan^{-1}[\sigma/(\omega \varepsilon)]}$$

$\sigma \ll \omega \varepsilon \quad \Rightarrow \quad$ lossless media

$$\beta \approx \omega \sqrt{\mu \varepsilon}, \quad \eta_c \approx \sqrt{\mu / \varepsilon}$$

$\sigma \gg \omega \varepsilon \quad \Rightarrow \quad$ perfect conductor

$$\eta_c \rightarrow 0 \text{ as } \sigma \rightarrow \infty$$

$$\text{loss tangent } \tan \delta = \frac{\sigma}{\omega \varepsilon}$$

good conductor $\tan \delta \gg 1$

poor conductor (good insulator) $\tan \delta \ll 1$

$$\hat{\varepsilon} = \varepsilon \left(1 - j \frac{\sigma}{\omega \varepsilon}\right) = \varepsilon' - j \varepsilon''$$

for a good dielectric $\varepsilon'' \ll \varepsilon'$

$$\text{loss tangent } \tan \delta = \frac{\varepsilon''}{\varepsilon'} \ll 1$$

$$\hat{k} = \omega \sqrt{\mu \hat{\varepsilon}} = \omega \sqrt{\mu (\varepsilon' - j \varepsilon'')} = \omega \sqrt{\mu \varepsilon' \left(1 - j \frac{\varepsilon''}{\varepsilon'}\right)} = \omega \sqrt{\mu \varepsilon'} \left(1 - j \frac{\varepsilon''}{\varepsilon'}\right)^{1/2}$$

$$\cong \omega \sqrt{\mu \varepsilon'} \left[1 + \frac{1}{2} \left(-j \frac{\varepsilon''}{\varepsilon'}\right) + \frac{1}{2} \left(\frac{1}{2} - 1\right) \frac{1}{2!} \left(-j \frac{\varepsilon''}{\varepsilon'}\right)^2 + \dots \right]$$

$$\cong \omega \sqrt{\mu \varepsilon'} \left[1 - j \frac{\varepsilon''}{2 \varepsilon'} + \frac{1}{8} \left(\frac{\varepsilon''}{\varepsilon'}\right)^2 + \dots \right]$$

$$\cong \omega \sqrt{\mu \varepsilon'} \left[1 + \frac{1}{8} \left(\frac{\varepsilon''}{\varepsilon'}\right)^2 \right] - j \omega \sqrt{\mu \varepsilon'} \left(\frac{\varepsilon''}{2 \varepsilon'}\right)$$

$$\because \frac{\varepsilon''}{\varepsilon'} \ll 1$$

$$\gamma = jk = \alpha + j\beta$$

$$\text{attenuation constant } \alpha \approx \omega \sqrt{\mu \varepsilon'} \left(\frac{\varepsilon''}{2 \varepsilon'}\right) \quad [np/m]$$

phase constant $\beta \approx \omega\sqrt{\mu\varepsilon'} \left[1 + \frac{1}{8} \left(\frac{\varepsilon''}{\varepsilon'} \right)^2 \right]$ [rad / m]

phase velocity $v_p = \frac{\omega}{\beta} \cong \frac{1}{\sqrt{\mu\varepsilon'}} \left[1 - \frac{1}{8} \right]$ [m / s]

intrinsic impedance $\hat{\eta}_c = \sqrt{\frac{\mu}{\hat{\varepsilon}}} = \sqrt{\frac{\mu}{\varepsilon' \left(1 - j \frac{\varepsilon''}{\varepsilon'} \right)}} = \sqrt{\frac{\mu}{\varepsilon'} \left(1 - j \frac{\varepsilon''}{\varepsilon'} \right)^{-1/2}}$

$$\cong \sqrt{\frac{\mu}{\varepsilon'} \left[1 + j \frac{\varepsilon''}{2\varepsilon'} - \frac{3}{8} \left(\frac{\varepsilon''}{\varepsilon'} \right)^2 + \dots \right]}$$

$$\approx \sqrt{\frac{\mu}{\varepsilon'} \left[1 - \frac{3}{8} \left(\frac{\varepsilon''}{\varepsilon'} \right)^2 \right]} + j \sqrt{\frac{\mu}{\varepsilon'} \left(\frac{\varepsilon''}{2\varepsilon'} \right)} \quad [\Omega]$$

for a good conductor $\sigma \gg \omega\varepsilon$

loss tangent $\tan \delta = \frac{\sigma}{\omega\varepsilon} \gg 1$

$$\hat{k} = \omega\sqrt{\mu\hat{\varepsilon}} = \omega\sqrt{\mu\varepsilon \left(1 - j \frac{\sigma}{\omega\varepsilon} \right)} \cong \omega\sqrt{\mu\varepsilon \left(-j \frac{\sigma}{\omega\varepsilon} \right)}$$

$$\cong \sqrt{j\omega\mu\sigma} = \sqrt{\omega\mu\sigma} e^{j\pi/4}$$

$$= \sqrt{\omega\mu\sigma} \left(\frac{1-j}{\sqrt{2}} \right)$$

$$\gamma = j\hat{k} = (1+j)\sqrt{\pi f\mu\sigma} = \alpha + j\beta$$

attenuation constant $\alpha \cong \sqrt{\pi f\mu\sigma}$

phase constant $\beta \approx \sqrt{\pi f\mu\sigma}$

intrinsic impedance

$$\hat{\eta}_c = \sqrt{\frac{\mu}{\hat{\varepsilon}}} = \sqrt{\frac{\mu}{\varepsilon \left(1 - j \frac{\sigma}{\omega\varepsilon} \right)}} \cong \sqrt{\frac{j\omega\mu}{\sigma}}$$

$$= \sqrt{\frac{\omega\mu}{\sigma}} e^{j\pi/4} = \frac{(1+j)}{\sqrt{2}} \sqrt{\frac{2\pi f\mu}{\sigma}} = (1+j)\sqrt{\frac{\pi f\mu}{\sigma}}$$

$$\hat{\eta}_c = (1+j)\sqrt{\frac{\pi f\mu}{\sigma}} = (1+j)(\sigma\delta)^{-1}$$

$$\text{skin depth } \delta = \frac{1}{\alpha} = \frac{1}{\sqrt{\pi f \mu \sigma}}$$

radiation pressure and electromagnetic moment